

Enlightened PLL Measurements Using Newly-Available Real-Time Spectrum Analysis

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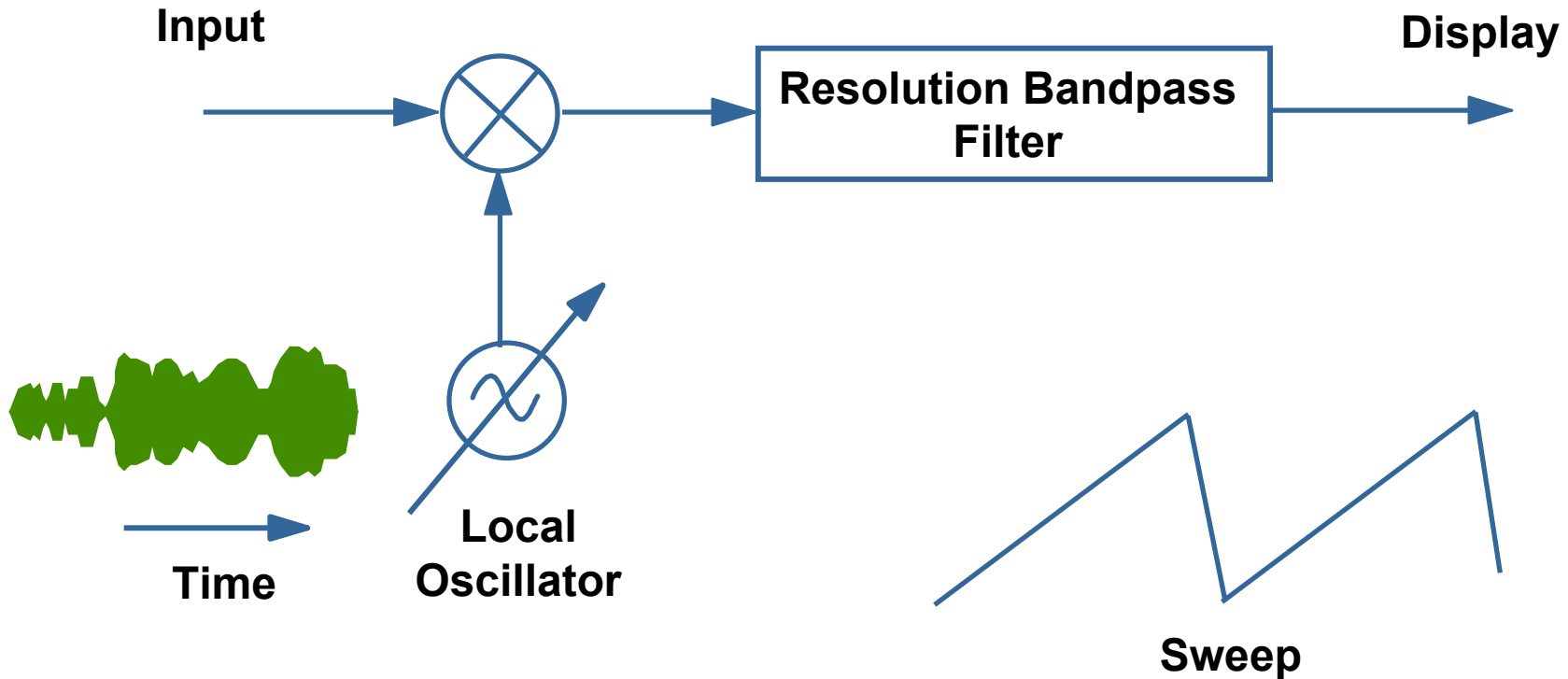
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Outline

- ▶ Overview of Swept and Real-Time Spectrum Analyzers
- ▶ Overview of Phase Locked Loops
- ▶ Measurements of Phase Locked Loops
- ▶ Live Demonstration of PLL Measurements Using Real-Time Spectrum Analysis
- ▶ Summary and Conclusions

Overview of Swept and Real-Time Spectrum Analyzers

Traditional Swept Spectrum Analyzer Architecture



Limitations of Traditional Tools

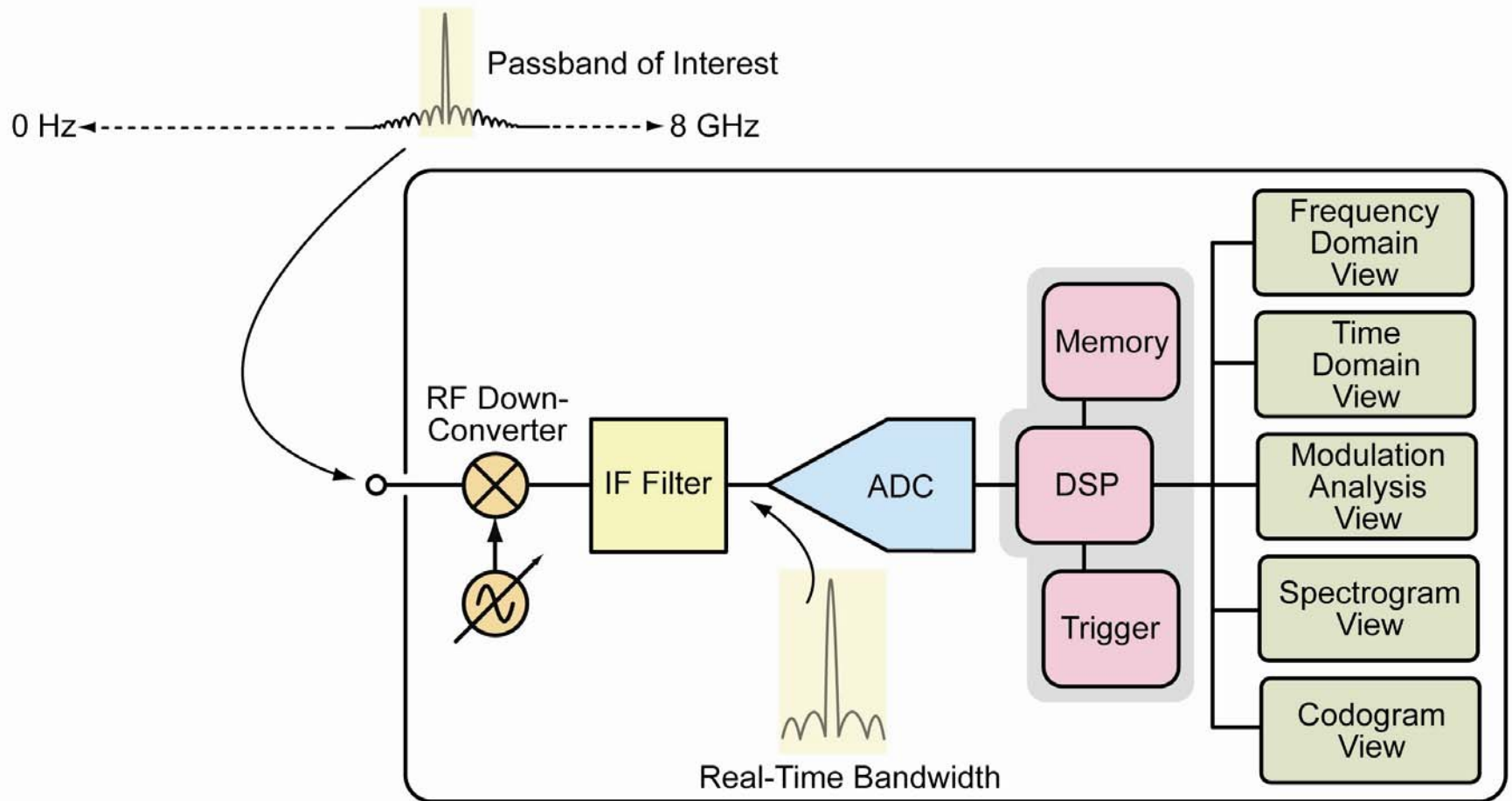
Traditional Swept Spectrum Analyzer Shortcomings

- ▶ Limited digital modulation analysis
- ▶ Limited triggering capabilities
- ▶ No time-history memory
- ▶ Single domain views
- ▶ Misses frequency changes over time

Traditional Vector Signal Analyzer (VSA) Shortcomings

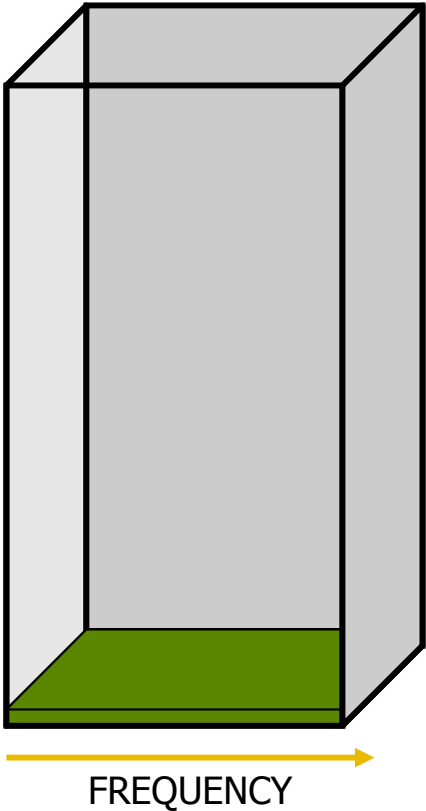
- ▶ Poor RF performance
- ▶ Limited triggering capabilities
- ▶ Limited correlated views
- ▶ Misses frequency and modulation changes over time

Real-Time Spectrum Analyzer Architecture

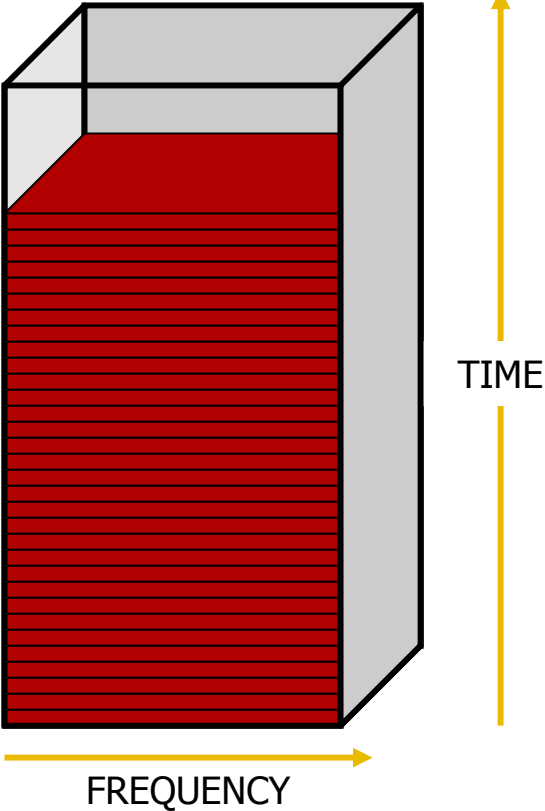


Swept vs. Real-time Seamless Capture

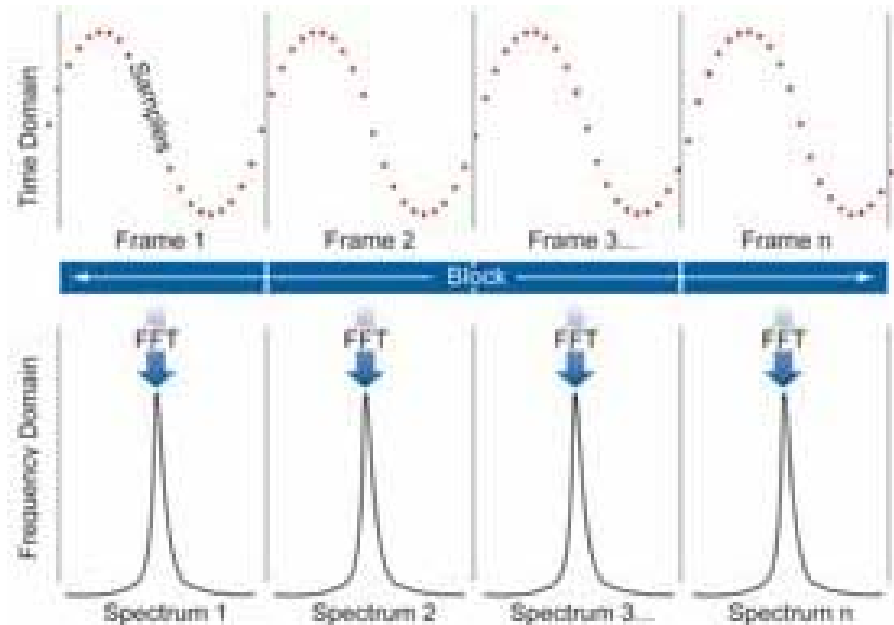
Swept Analyzers



Real-Time Analyzers



The FFT Transformation of the Signal



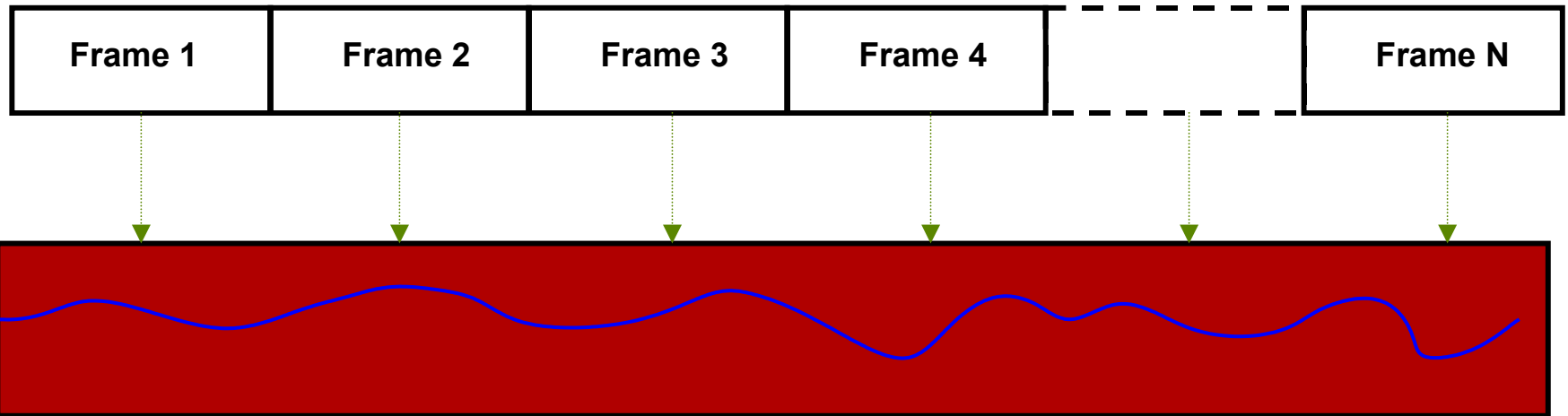
- ▶ Frame after frame, the time domain is converted to the frequency domain with the FFT

Real-Time SA

Trigger on an RF signal based on voltage, power, or frequency characteristics



Capture the signal seamlessly in time, store it frame by frame into memory



TIME

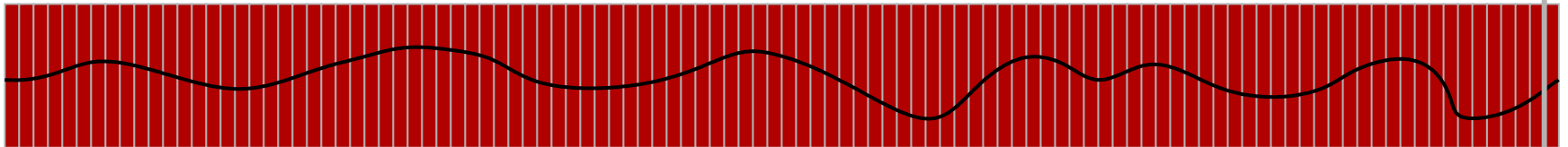


Real-Time SA – View of Modulation Domain

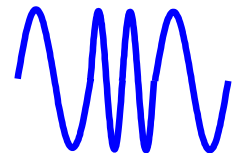
Trigger | on an RF signal based on power or frequency characteristics



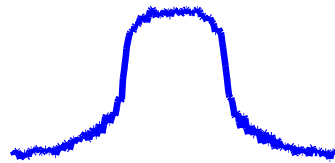
Capture | the signal seamlessly in time, into memory



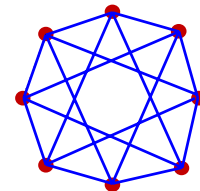
Analyze | the signal simultaneously in multiple domains



Time Domain

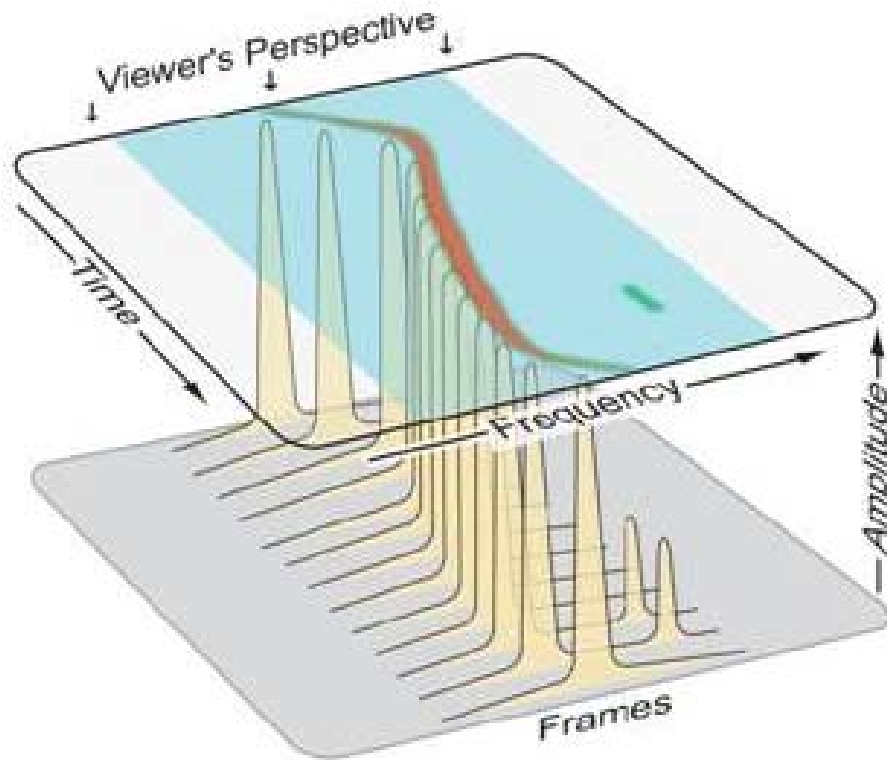


Frequency Domain



Modulation Domain

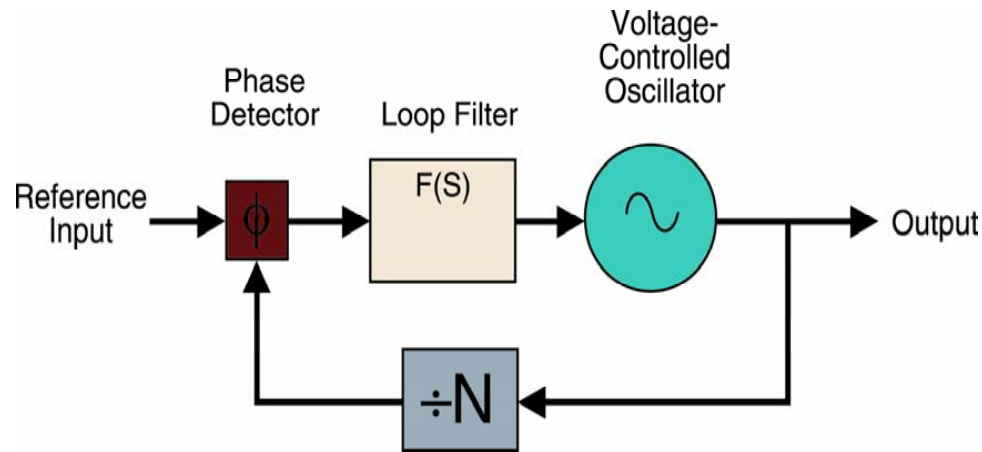
Seamless Capture and Spectrogram



- ▶ The spectrogram shows how an RF signal changes over time in the frequency domain
- ▶ Frequency is the horizontal axis, time is the vertical axis, and power is represented by the color of the trace

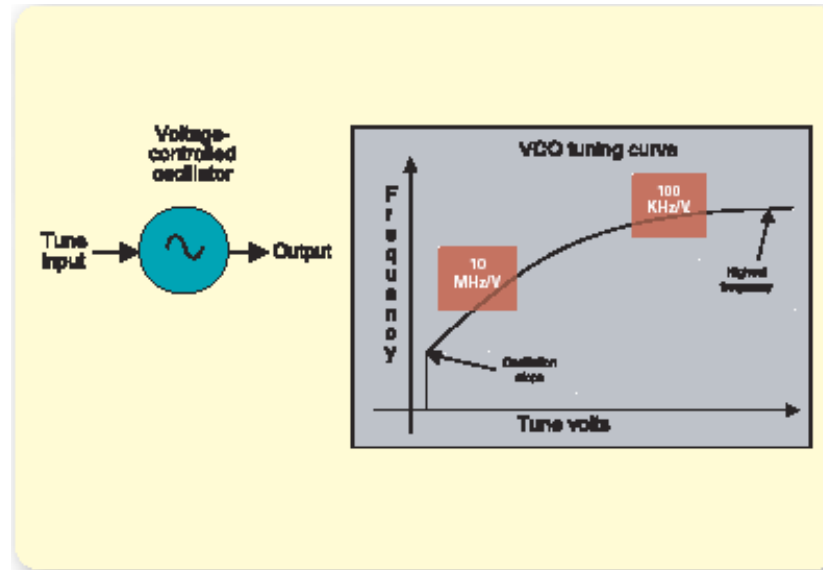
Overview of Phased Locked Loops (PLLs)

The Building Blocks



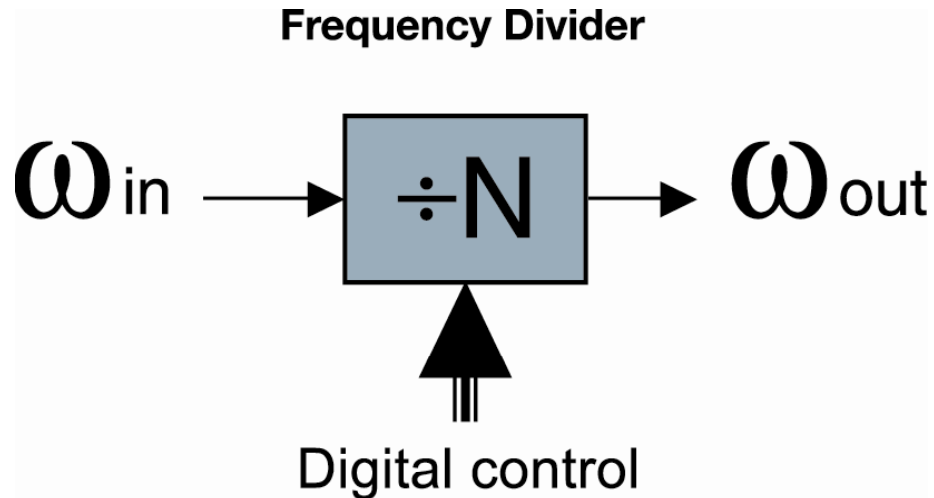
- ▶ PLL locks the phase of the output to N times the phase of the input
- ▶ PLLS behave linearly in the vicinity of lock
- ▶ Non-linear effects dominate when PLL is unlocked

Voltage Controlled Oscillator (VCO)



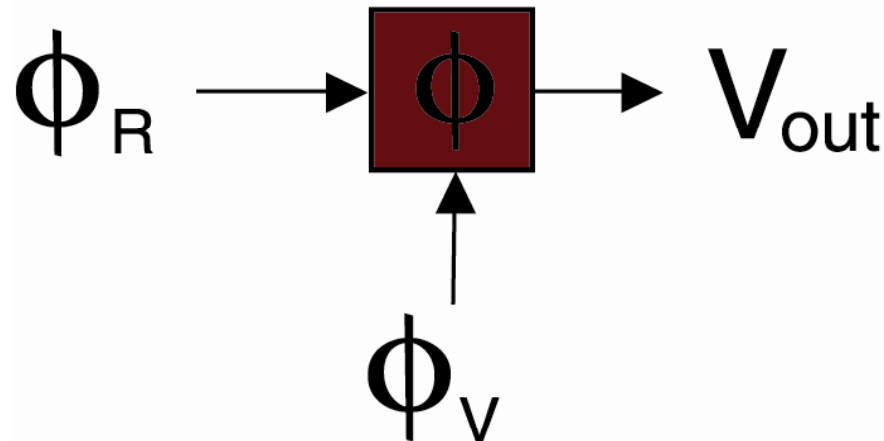
- ▶ VCO output signal is controlled through a tuning signal
- ▶ Tuning signal is typically a voltage, but it can be another variable (current, numerically, etc)
- ▶ VCOs are often non-linear, having tuning sensitivities that vary greatly
- ▶ Other non linear behaviors include:
 - Min/Max tune frequencies
 - Tune voltages where oscillations stop

Frequency Divider



- ▶ Typically a digital counter produces one output for every N inputs
- ▶ Both Phase and Frequency are scaled by the ratio N
- ▶ N is generally an integer, however Fractional-N techniques exist to provide ratios that are not whole numbers

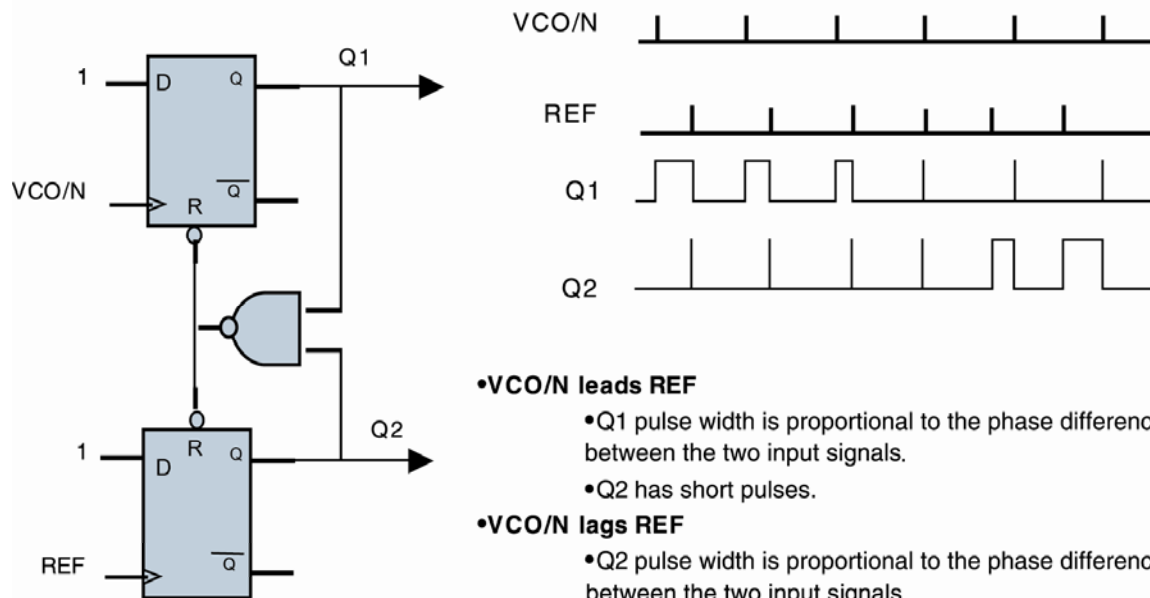
Phase Detector



- ▶ Produces an output that is proportional to the phase difference between two inputs signals
- ▶ Φ_R = phase of the reference signal
- ▶ Φ_V = phase of the VCO signal
- ▶ V_{out} = a signal that is proportional to the difference between Φ_R and Φ_V

Phase Frequency Detector

Phase Frequency Detector Logic and Timing



- **VCO/N leads REF**

- Q1 pulse width is proportional to the phase difference between the two input signals.
- Q2 has short pulses.

- **VCO/N lags REF**

- Q2 pulse width is proportional to the phase difference between the two input signals
- Q1 has short pulses.

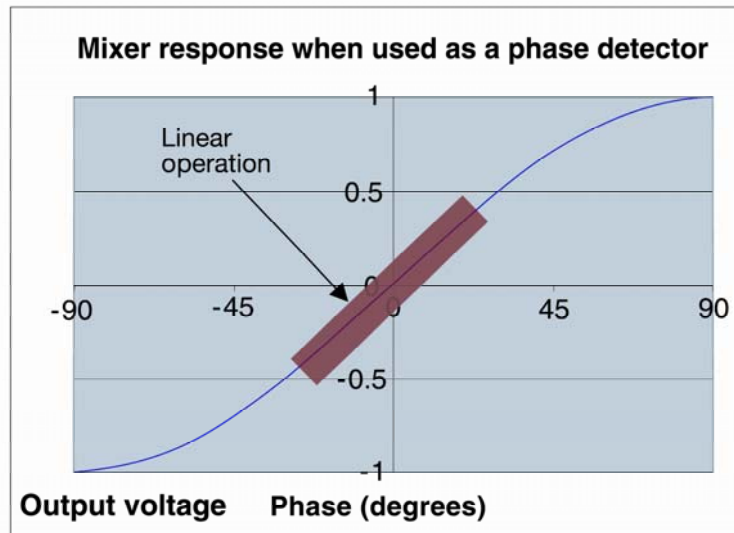
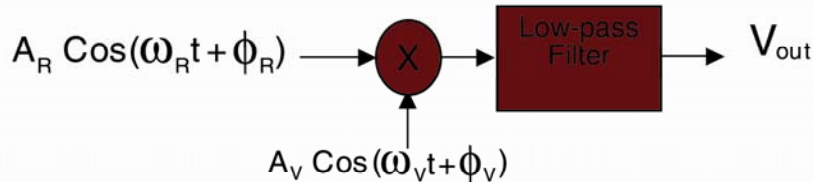
- **VCO/N has a higher frequency than REF: Frequency steering down.**

- Q1 is high except for occasional short glitches. Q2 is low except for occasional glitches.

- **VCO/N has a lower frequency than REF: Frequency steering up.**

- Q2 is high except for occasional short glitches. Q1 is low except for occasional glitches.

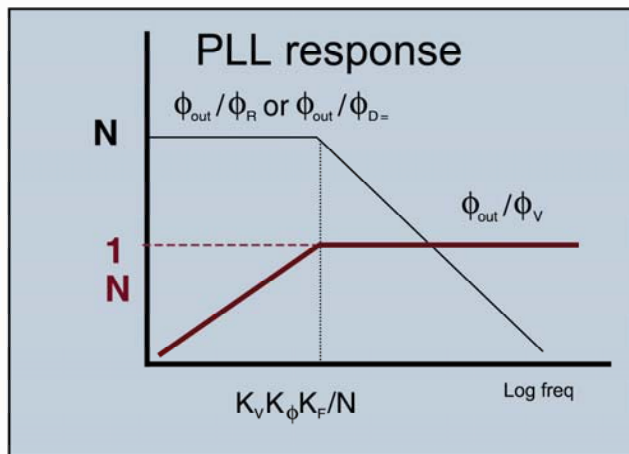
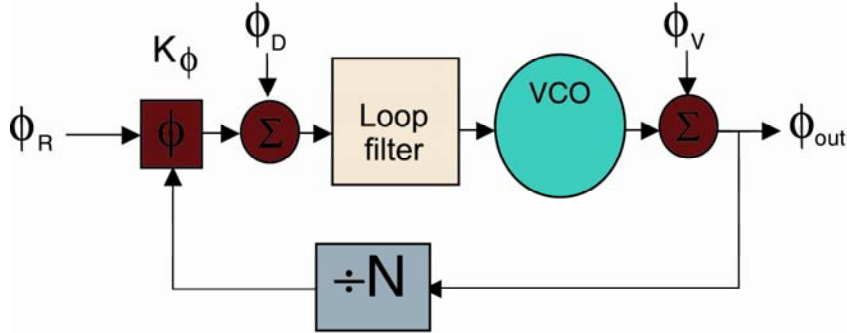
Mixer Phase Detector



- ▶ Multipliers or Mixers can be used as phase detectors
- ▶ The output of the mixer is proportional to the sine of the phase difference between the inputs
- ▶ $V_{out} = K_{det} A_r A_v \cos(\Phi_R - \Phi_V)$
- ▶ K_{det} = mixer conversion gain
- ▶ A_r and A_v are the amplitudes of the input signals
- ▶ Φ_R and Φ_V are the phases of the two input signals

Loop Filter

Simplified phase locked loop linear analysis

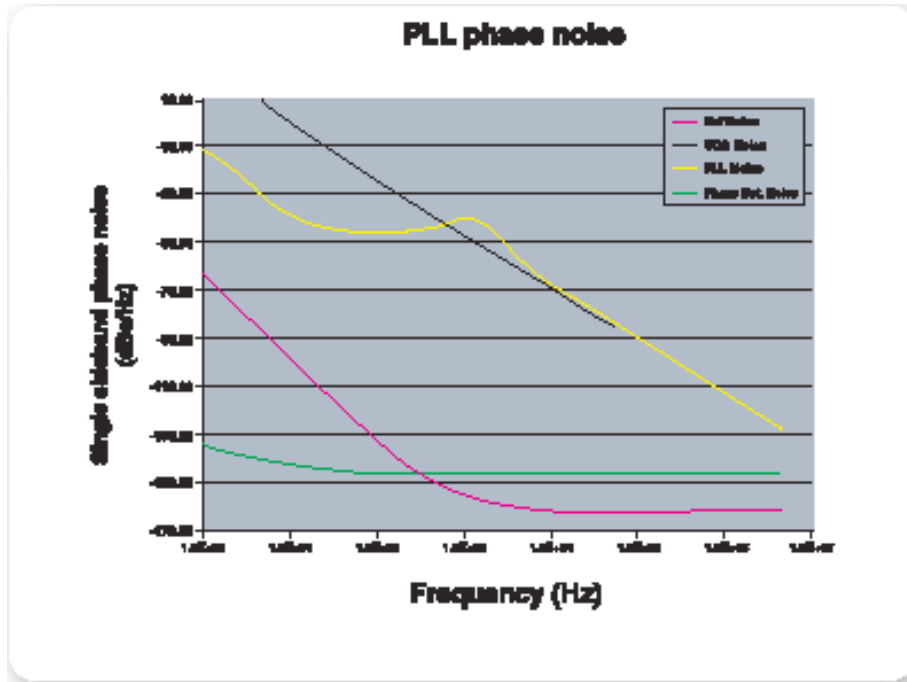


- ▶ Loop Filters are used to tailor the PLL response to optimize:

- Bandwidth
- Switching speed
- Settling time
- Spurious levels, etc

- ▶ Loop filters often include integrators and therefore are subject to the same linearity concerns as other active devices

Linear Operation/Response

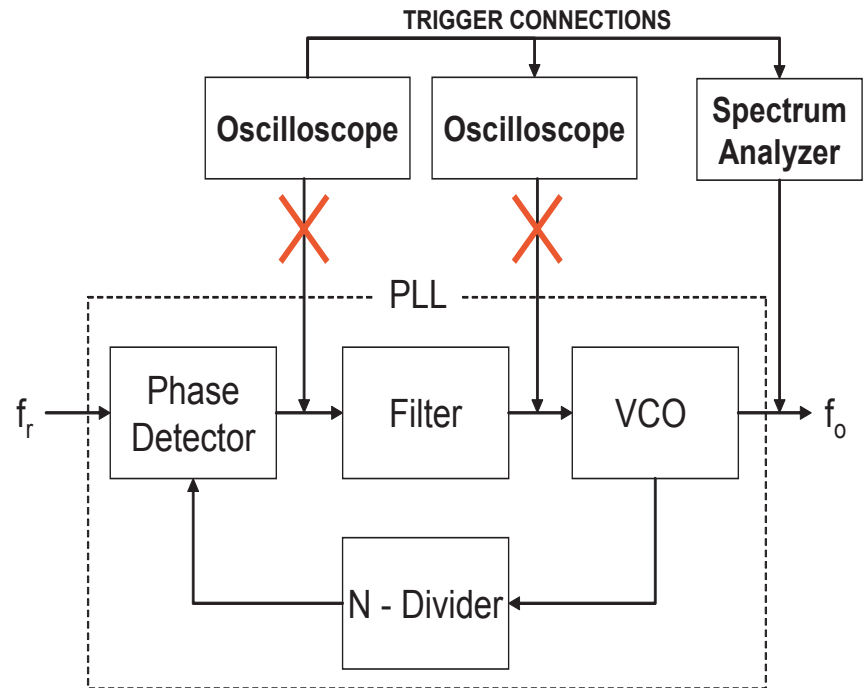


- ▶ All oscillators exhibit phase fluctuations
- ▶ PLLs can be used to improve the phase stability of oscillators
- ▶ The output of the PLL varies due to fluctuations in the reference, VCO and the signal applied to the Phase Detector
 - Phase Noise at low frequencies is dominated the reference signal and the signal applied to the phase detector
 - The VCO dominates the phase noise performance at higher frequency

Measurements of Phased Locked Loops (PLLs)

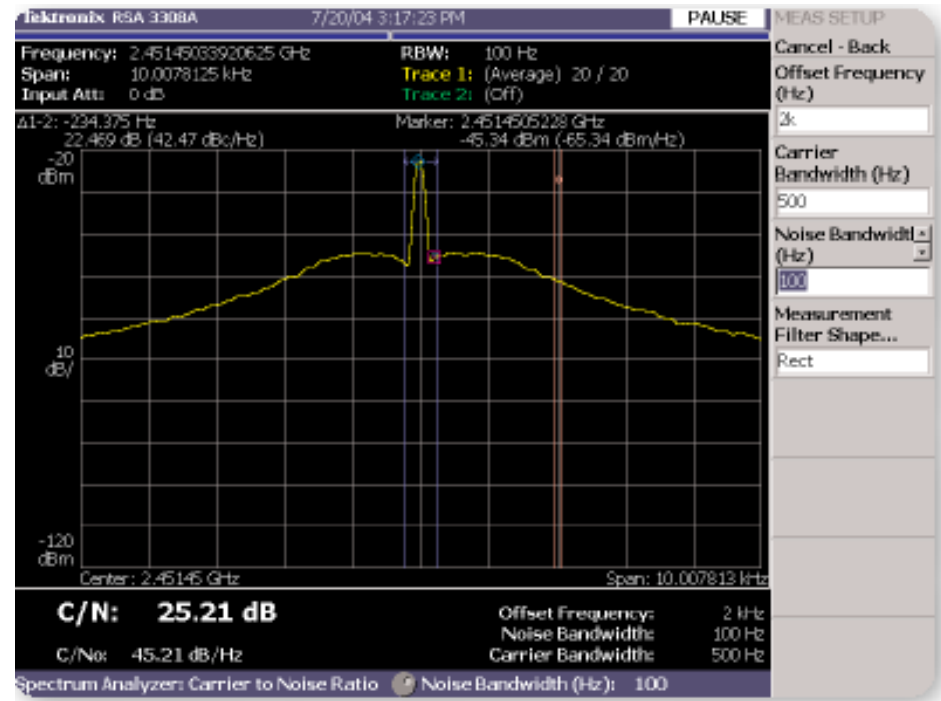
Limitations of Traditional Test Methods

- ▶ Traditional methods relied on sampling the signal at each node
- ▶ Higher levels of circuit integration limits access to individual nodes
- ▶ Traditional test equipment requires complex triggering arrangements
- ▶ Traditional methods do not provide time correlation between frequency and time domain



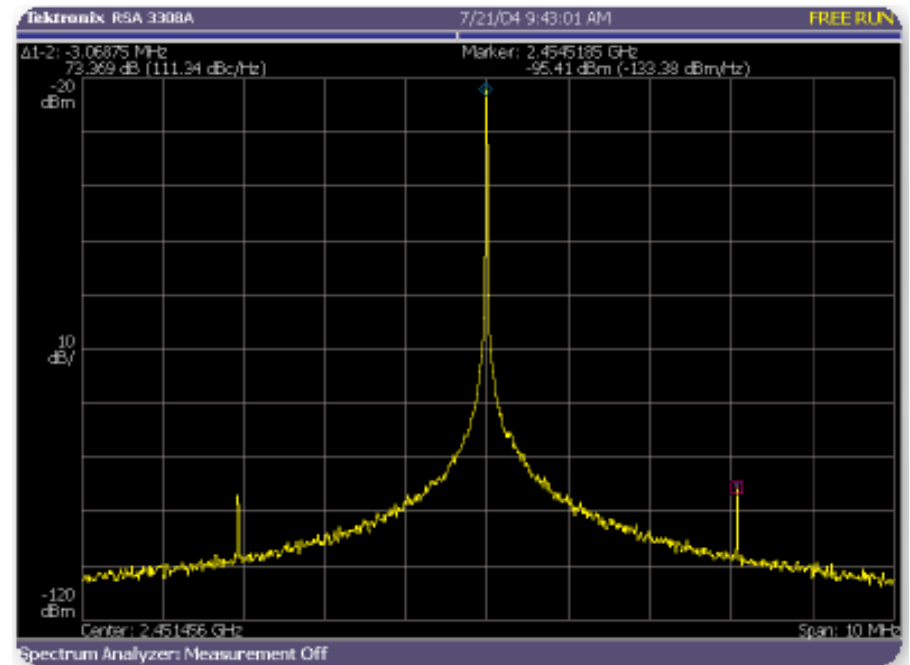
Phase Noise Measurements

- ▶ Phase Noise typically expressed as a ratio of sideband power in 1Hz BW to the signal power (dBc/Hz)
- ▶ Phase Noise is measured at specific offsets from the carrier frequency
- ▶ Measurement Example:
 - CF = 2.4515 GHz
 - Offset = 2 kHz
 - Noise Bandwidth = 100 Hz
 - C/N = 25.21 dB
 - C/No = 45.21 dB
- ▶ This traditional view only shows a single snapshot in time

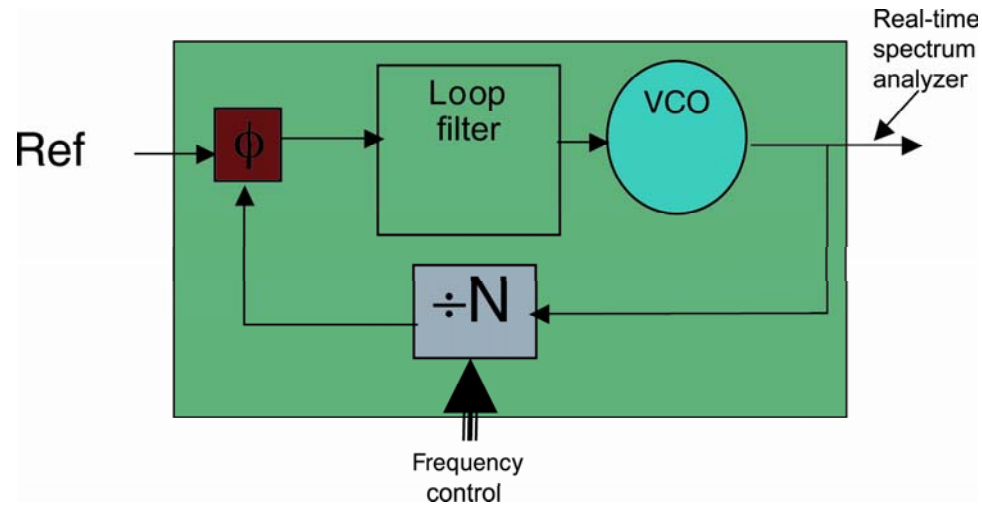


Spurious Signals

- ▶ There are many sources of spurious signals that can affect the output of a PLL
 - Switching circuitry
 - Power supply switching
 - Close by oscillators
 - IMD
- ▶ The reference signal can also generate spurious
 - These spurious signals are always at multiples of the reference signal
- ▶ The loop filter can be used to reduce spurious, but in the real world, they can never be eliminated
- ▶ This traditional view only shows a single snapshot in time



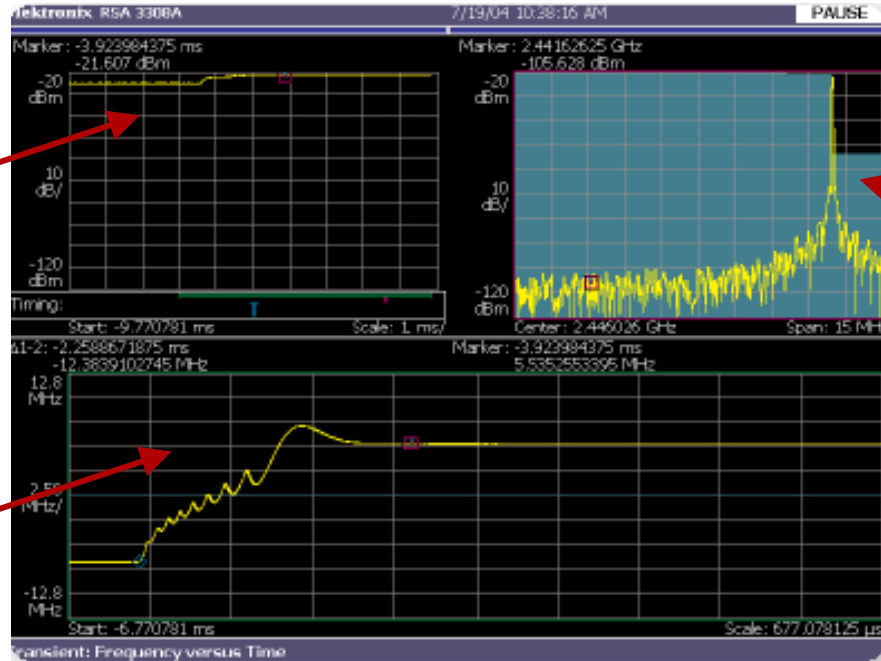
Non Linear Behavior



- ▶ During switching PLLs experience a lot non-linear behavior until the phase detector can close the loop
- ▶ Measurement of the switching transients requires time domain analysis
- ▶ Traditional methods required probing at each stage of the PLL, however this is not always possible with integrated PLL circuits

Capturing Transient Events, Using Real-Time SA

RF Output Power vs. time



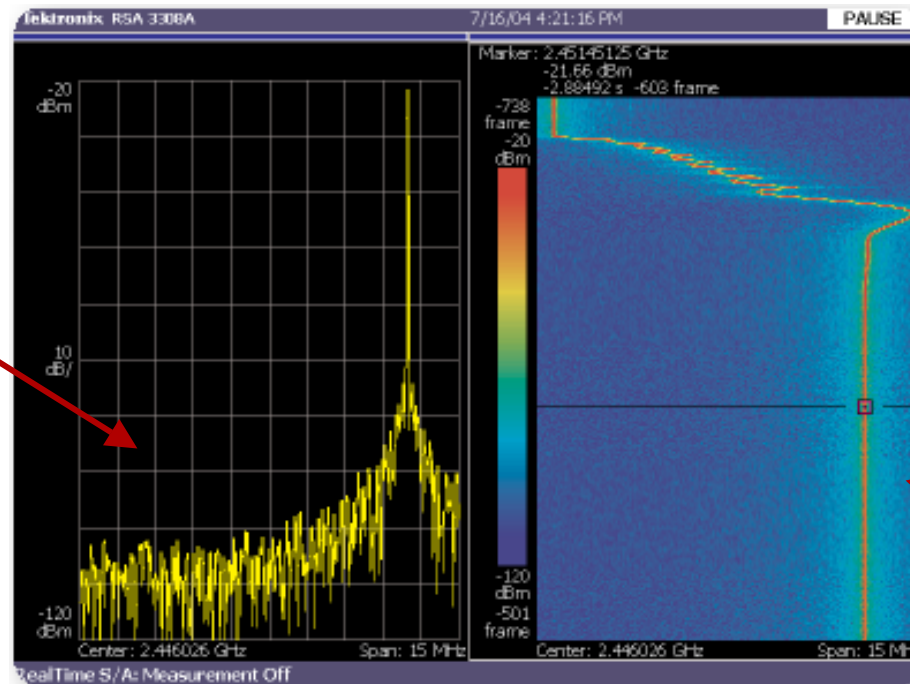
Traditional Amplitude vs. Frequency

RF Output Frequency vs. time

- ▶ Real-Time Spectrum Analyzers can capture the transient event by triggering the instant that the output of the PLL changes
- ▶ Capturing a seamless history of the signal enables the non-linear behavior to be analyzed in both the time and frequency domains

Frequency Changes over Time, Using Real-Time SA

**Traditional
Amplitude
vs.
Frequency**

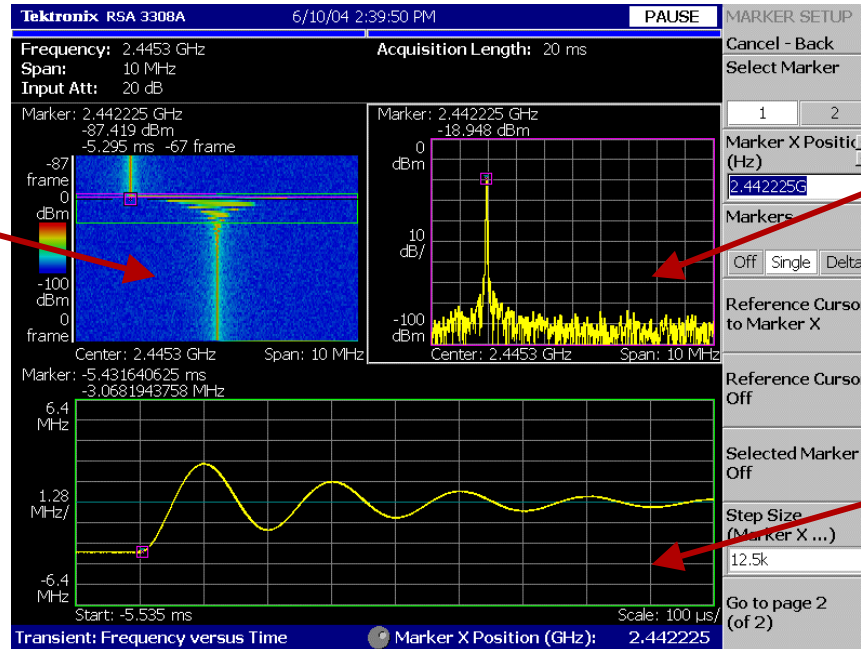


**Frequency
variation
over time**

- ▶ The dynamic frequency behavior of a PLL determines its spectral output
 - Understanding the spectral output before, during and after the transient event is important for designers
 - To minimize interference spectral mask requirements typically have to be met at all times
- ▶ Viewing frequency changes over time is critical for designers of PLLs

Multiple Domain View, Using Real-Time SA

Transient response of step frequency change



Traditional Amplitude vs. Frequency

Frequency vs. time (settling time)

- ▶ Trigger the moment the PLL loses lock
- ▶ Real-Time seamless capture ensures no data is missing
- ▶ Analyze the single shot event completely with time correlated multi-domain views Increased levels of integration makes the analysis of linear and non linear behavior critical
- ▶ Analysis of the transient events in the time, frequency and modulation domains

Live Demonstration of PLL Measurements Using Real-Time Spectrum Analysis

Demonstration Prelude

- ▶ The PLL is one of the most common circuits in use today
- ▶ Technology trends in PLL development are leading to complex designs:
 - Increasing output frequencies
 - Higher output frequency stability
 - Faster settling time
 - Faster switching speeds
- ▶ Commercial demands are leading to the creation of smaller, lower cost PLL
 - PLLs are becoming more integration and are often just a “black box”

Demonstration Measurements

- ▶ Several different measurements must be made on a single shot event:
 - Frequency switching speed (Time Domain)
 - Frequency settling behavior (Time Domain)
 - Spectral purity/splatter (Frequency Domain)
 - Transient instabilities

- ▶ Repeatability measurements must be made on short, very frequent events
 - Output Frequency
 - Frequency settling time
 - Spectral purity
 - Spectral splatter

Summary & Conclusions

Summary & Conclusions

- ▶ Modern PLL circuits have limited access for monitoring
- ▶ Characterizing PLL's over both time and frequency provides the greatest insight to the circuits behavior
- ▶ Settling time in frequency agile systems is very critical for proper timing.
- ▶ The tools must be up to the tasks for timely and accurate results.
- ▶ Proper signal capturing and measurement are often more than half the battle.